

# Performance Evaluation of Conventional Inverters Driven PMSM Drive using Microcontroller

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## Abstract:

This paper focues on Performance evaluation of conventional inverters driven PMSM drive using microcontroller. The purpose of this paper is to decrease the ripples in torque of Micro controllers based PMSM drive. The innovative method consists of conventional inverters, switched-mode power supply (SMPS), PMSM motor and Microcontroller. It is used to maximize fundamental component of torque also. The both the results of the suggested Micro controller are compared on the basis of torque and speed and improves that reduces the torque ripple and improve the dynamic response of the system in comparative analysis. This paper organizes introduction, mathematical model of the PMSM, proposed Microcontroller technique, hardware and simulation results and conclusion in different sections.

*Keywords*: Switched-mode power supply (SMPS), conventional inverters, Microcontroller, Permanent magnet synchronous motors (PMSM), EPROM, Clock generator.

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# I. INTRODUCTION

In industrial application, Electric motors (EMs) and generators are the primary workhorses. The generators convert energy transformation to charge the batteries for motor operation. Required torque developed by motors for drive the wheek. The permanent magnets, switched reluctance, and induction motors are used in industrial application [1-4]. This paper explains Simulation and Hardware Implementation of VSI -Fed PMSM drives for automotive application. For the speed &torque control,VSI fed PMSM drive is one of the method used. The VSI fed PMSM drive system is designed, simulated and implemented using Microcontroller. A hardware setup is designed and implemented based on a Four-pole 0.25 HP PMSM. The control hardware consists of, Main power circuit, Control circuit, Isolator and driver circuit and PMSM [5-11]. The fixed dc voltage is obtained from three phase rectifier circuit and uses shunt capacitor for filter purpose. The output of three phase rectifier circuit is given to MOSFET bridge inverters in power circuit [12].The function of control circuit is to tum the MOSFET by using Gating pulses Isolator and driver circuit is used for proper isolation by using opt isolator [25-26]. The steps involved in any drive system are design, building its hardware and its testing. Once the system is designed, depending on designed values of various components, the system hardware is

developed. Once the assembly of hardware is completed then it is tested by carrying out various experiments on the developed system to test whether the system is working satisfactory or not. If it is not working properly then the error must be removed and the system must be updated and modified. The drive developed here is operating variable frequency mode. The theme of the paper is Simulation and Hardware analysis of torque ripple minimization of PMSM drive for automotive application.

## II. MATHEMATICAL MODEL OF PMSM

## A. PMSM modeling in rotor reference frame

There are two stator windings in the dq reference frame and field winding is absent in rotor of PMSM. The directaxis winding is along with the axis of magnetic pole. If PMSM running in anti-clockwise direction, at the speed of ' $\omega$ r',

The induced voltage in the direct-axis winding:

$$u_d = R_d i_d + \frac{d\lambda_d}{dt} - \omega_r \lambda_q \tag{1}$$

*Where*, *'id'* and *'Rd'* are called direct-axis stator current and resistance respectively

The induced voltage in the quadrature-axis winding:

$$u_q = R_q i_q + \frac{d\lambda_q}{dt} - \omega_r \lambda_d \tag{2}$$

Where, 'Rq' and 'iq' are called the quadrature -axis resistance and current of stator.

$$\lambda_d = L_d i_d + \lambda_m \tag{3}$$

 $\lambda d = flux$ -linkage in the direct-axis of stator in webers,  $\lambda m$  is the PM rotor flux

$$\lambda_q = \mathbf{L}_q \mathbf{i}_q \tag{4}$$

 $\lambda q = flux$ -linkage in the quadrature-axis stator (Wb)

In this case of quadrature-axis, there are no magnets so  $\lambda m$  is absent.

Considering round rotor PMSM, we have  $\mathbf{L} = \mathbf{L}$ 

$$L_d - L_q$$
(5)

The PMSM torque equation is given by:

$$T_e = \frac{3 p}{2 2} (\lambda_d i_q - \lambda_q i_d)$$
(6)

Substituting for  $\lambda d'$  and  $\lambda q'$  in the torque equation of PMSM,

$$T_e = \frac{3 p}{2 2} \left[ (\lambda_d i_d + \lambda_m) i_q - \mathcal{L}_q i_q i_d \right]$$
(7)

$$T_e = \frac{3 p}{2 2} \left[ (L_d - L_q) i_d i_q + \lambda_m i_q \right]$$
(8)

The two components of torque developed are:

Reluctance torque 
$$=\frac{3}{2}\frac{p}{2}(L_{d}-L_{q})i_{d}i_{q}$$
 (9)

field torque 
$$=\frac{3}{2}\frac{p}{2}$$
  $\lambda_m i_q$  (10)  
 $T_e = \frac{3}{2}\frac{p}{2}$   $\lambda_m i_q$  (11)

In a round rotor PMSM the electromagnetic torque present is the field torque present due to the PM flux linkage, 
$$\lambda m$$
. For a chosen PMSM, the PM rotor flux-linkage ( $\lambda m$ ) and the number of poles ( $p$ ) is constant. Hence, for the round-rotor PMSM, the electromagnetic torque equation is

$$T_e = Kt i_q \tag{12}$$

Where,  $Kt = Torque \ constant$  $K_t = \frac{3}{2} \frac{p}{2} \quad \lambda_m$ 

Therefore, electro-magnetic torque is given by

$$T_e = T_l + B \quad \omega_m + J \frac{d\omega_m}{dt} \quad (14)$$

 $\omega m = \text{rotor's mechanical speed}, \quad \omega r = \text{rotor's electrical speed}$ 

# III. SIMULATION RESULT

The simulation response of Voltage source inverterfed PMSM drives at different speed range is shown in figure 5 to figure 9.



Figure 1.Simulation diagram



Figure 2.D-axis &Q-axis Currents model



Figure 3.PM Synchronous Motor model



Figure 4. PMSM Characteristics diagram

(13)







Figure 5.Speed, torque and Current response at 1000 to 1800 rpm



Figure 6.Speed, torque and Current response at 1200 to 1800 rpm



Figure 7.Speed, torque and Current response at 1400 to 1800 rpm



Figure 8.Speed ,torque and Current response at 1600 to 1800 rpm



Figure 9.Speed, torque and Current response at 1800 rpm

## IV. HARDWARE RESULT ANALYSIS

Initially the drive operation was worked on dummy load (3 phase star connected lamps) by connecting the motor to the drive. Various outputs are taken out from the control card on the drive. Firstly the frequency knob on the drive was kept at its minimum position. The converter card was then switched on. The power supply status was indicated by all the LEDs inside the unit. By using CRO probes, the waveforms at various test points were observed.



Figure 10.Block diagram of conventional inverters driven PMSM

While observing the waveforms on CRO, care was taken to use the unearthed CRO only. The basic timer IC 555 output frequency was observed at the test point. This frequency can be changed by changing the position of frequency knob. The pulses for various gate drives were observed at the test points. The output frequency was observed at the test point. The minimum and maximum frequency was noted and tabulated in the result table. The frequency knob position was slowly changed along with the set speed pot. Slowly the lamps were turned on as the soft start in the result table. The output line and phase voltage waveforms for the three phases were observed on the CRO at the corresponding test points. After observing the satisfactory performance of the control card then the motor was connected to the drive. The dummy lamp load was now removed and the motor was connected in the output connector. Thus the motor runs at variable speed depending on the frequency. After testing the various test points the dummy star connected load is removed and across the output of the bridge inverter PMSM is connected. Then gradually increasing the load on the shaft of the motor for different values of frequency speed of the motor is measured.



TABLE NO.1: MOTOR SPEED VARIATION BY FREQUENCY

Sr. No	Time (m.s.)	Frequency (Hz)	Actual Speed (rpm)	Measu red Speed (rpm)	Voltage (Volts)
1	30	33.4	995	1005	260
2	25	40.1	1250	1219	265
3	22	45.5	1462	1370	265
4	20	50.2	1530	1518	265
5	18	55.6	1765	1672	265
6	17	59.2	1785	1795	265

TABLE NO.2: MOTOR SPEED VARIATION BY LOAD AT 33.4

Sr.	Load	Frequency	Actual	Expected	Torque	Output
No.	(gm)	(Hz)	Speed	Speed	(N-m)	Power
			(rpm)	(rpm)		( W)
1	500	33.4	1110	995	0.15715	16.59
2	1000	33.4	1110	995	0.2743	35.28
3	1500	33.4	1110	995	0.4814	48.99
4	2000	33.4	1110	995	0.6186	65.45
5	2500	33.4	1110	995	0.8357	79.98
6	3000	33.4	1110	995	0.9829	99.49

TABLE NO.3: MOTOR SPEED VARIATION BY LOAD AT 50.2HZ

Sr.	Load	Frequency	Actual	Expected	Torque	Output
No.	(gm)	(Hz)	Speed	Speed	(N-m)	Power
			(rpm)	(rpm)		( W)
1	500	50.2	1500	1540	0.16715	25.89
2	1000	50.2	1500	1540	0.3043	49.96
3	1500	50.2	1500	1540	0.4814	76.45
4	2000	50.2	1500	1540	0.6886	99.79
5	2500	50.2	1500	1540	0.7857	125.51
6	3000	50.2	1500	1540	0.9229	148.78

TABLE NO.4: MOTOR SPEED VARIATION BY LOAD AT 59.2 HZ

Sr. No	Load (gm)	Frequ ency (Hz)	Actual Speed (rpm)	Expecte d Speed (rpm)	Torque (N-m)	Output Power (W)
1	500	59.2	1700	1770	0.1571	29.64
2	1000	59.2	1700	1770	0.3243	58.16
3	1500	59.2	1700	1770	0.4914	89.73
4	2000	59.2	1700	1770	0.6588	115.33
5	2500	59.2	1700	1770	0.7957	142.90
6	3000	59.2	1700	1770	0.9229	172.49



Figure 11.Speed- Frequency characteristics



Figure 12.Load -Speed Characteristics at 33.3 Hz



Figure 13.Load -Speed Characteristics at 50 Hz



Figure 14.Load -Speed Characteristics at 59 Hz

## CONCLUSION

This paper focuses on simulation and hardware analysis of conventional inverters driven PMSM drive using microcontroller. A detailed Simulink model of torque minimization of PMSM drive has being developed. It is observed that, at constant frequency, speed remains constant irrespective of load. The motor runs at synchronous speed. Speed also gets changed accordingly the inverter frequency .The overall motor performance can be very well judged from the performance characteristic shown. Finally we conclude simulation that results and hard ware implementation shall give better performance analysis. Hence it is used in automotive application.

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